

## LMV7219

### 7 nsec, 2.7V to 5V Comparator with Rail-to Rail Output

#### General Description

The LMV7219 is a low-power, high-speed comparator with internal hysteresis. The LMV7219 operating voltage ranges from 2.7V to 5V with push/pull rail-to-rail output. This device achieves a 7ns propagation delay while consuming only 1.1mA of supply current at 5V.

The LMV7219 inputs have a common mode voltage range that extends 200mV below ground, allowing ground sensing. The internal hysteresis ensures clean output transitions even with slow-moving input signals.

The LMV7219 is available in the SC70-5 and SOT23-5 packages, which are ideal for systems where small size and low power are critical.

#### Features

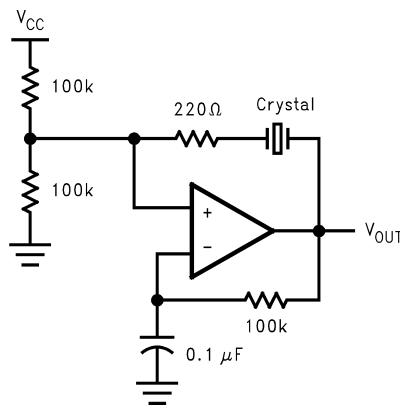
( $V_S = 5V$ ,  $T_A = 25^\circ C$ , Typical values unless specified)

- Propagation delay 7ns
- Low supply current 1.1mA
- Input common mode voltage range extends 200mV below ground
- Ideal for 2.7V and 5V single supply applications
- Internal hysteresis ensures clean switching
- Fast rise and fall time 1.3ns
- Available in space-saving packages:
  - 5-pin SC70-5
  - SOT23-5

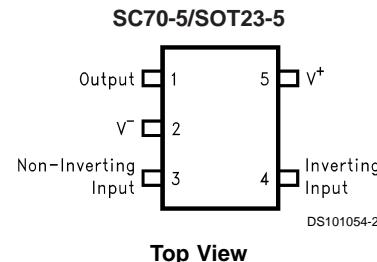
#### Applications

- Portable and battery-powered systems
- Scanners
- Set top boxes
- High speed differential line receiver
- Window comparators
- Zero-crossing detectors
- High-speed sampling circuits

#### Typical Application



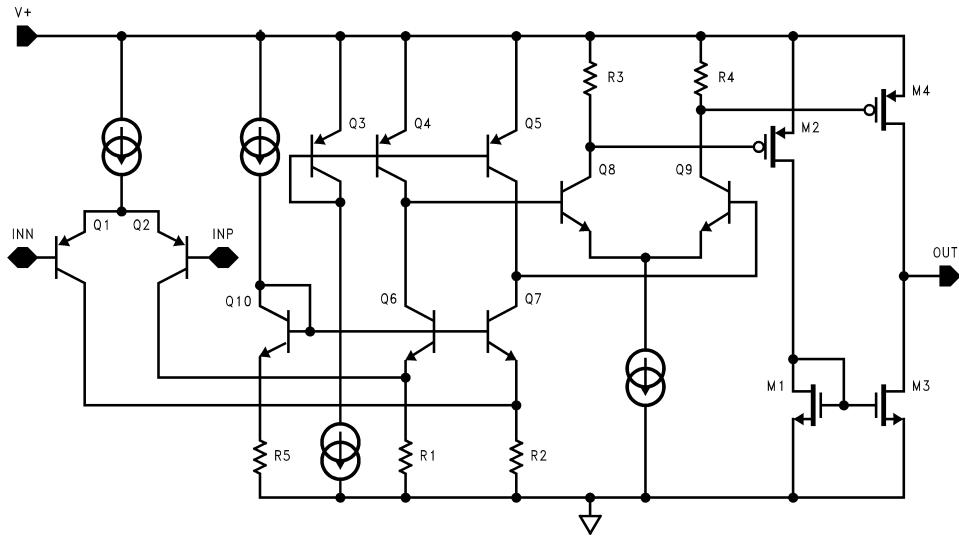
## Connection Diagram



## Ordering Information

Package	Part Number	Marking	Supplied as	NSC Drawing
5-pin SC70-5	LMV7219M7	C15	1k Units Tape and Reel	MAA05A
	LMV7219M7X	C15	3k Units Tape and Reel	
5-pin SOT23-5	LMV7219M5	C14A	1k Units Tape and Reel	MA05B
	LMV7219M5X	C14A	3k Units Tape and Reel	

## Simplified Schematic



## Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

ESD Tolerance (Note 2)		
Machine Body	150V	
Human Model Body	2000V	
Differential Input Voltage	$\pm$ Supply Voltage	
Output Short Circuit Duration	(Note 3)	
Supply Voltage ( $V^+ - V^-$ )	5.5V	
Soldering Information		
Infrared or Convection (20 sec)	235°C	
Wave Soldering (10 sec)	260°C (lead temp)	

Voltage at Input/Output pins

Current at Input Pin (Note 9)

## Operating Ratings

Supply voltages ( $V^+ - V^-$ )	2.7V to 5V
Junction temperature range (Note 4)	-40°C to +85°C
Storage Temperature Range	-65°C to +150°C
Package Thermal Resistance	
SC70-5	478°C/W
SOT23-5	265°C/W

## 2.7V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V_{CM} = V^+/2$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ ,  $C_L = 10\text{pF}$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Limit (Note 6)	Units
$V_{OS}$	Input Offset Voltage		1	6 <b>8</b>	mV max
$I_B$	Input Bias Current		450	950 <b>2000</b>	nA max
$I_{OS}$	Input Offset Current		50	200 <b>400</b>	nA max
CMRR	Common Mode Rejection Ratio	$0\text{V} < V_{CM} < 1.50\text{V}$	85	62 <b>55</b>	dB min
PSRR	Power Supply Rejection Ratio	$V^+ = 2.7\text{V}$ to $5\text{V}$	85	65 <b>55</b>	dB min
$V_{CM}$	Input Common-Voltage Range	CMRR > 50dB	$V_{CC} -1$	$V_{CC} -1.2$ $V_{CC} -1.3$	V min
			-0.2	-0.1 <b>0</b>	V max
$V_O$	Output Swing High	$I_L = 4\text{mA}$ , $V_{ID} = 500\text{mV}$	$V_{CC} -0.22$	$V_{CC} -0.3$ $V_{CC} -0.4$	V min
		$I_L = 0.4\text{mA}$ , $V_{ID} = 500\text{mV}$	$V_{CC} -0.02$	$V_{CC} -0.05$ $V_{CC} -0.15$	
	Output Swing Low	$I_L = -4\text{mA}$ , $V_{ID} = -500\text{mV}$	130	200 <b>300</b>	mV max
		$I_L = -0.4\text{mA}$ , $V_{ID} = -500\text{mV}$	15	50 <b>150</b>	
$I_{SC}$	Output Short Circuit Current	Sourcing, $V_O = 0\text{V}$ (Note 3)	20		mA
		Sinking, $V_O = 2.7\text{V}$ (Note 3)	20		
$I_S$	Supply Current	No load	0.9	1.6 <b>2.2</b>	mA max
$V_{HYST}$	Input Hysteresis Voltage	(Note 10)	7		mV
$V_{TRIP^+}$	Input Referred Positive Trip Point	(see Figure 1)	3	8	mV max
$V_{TRIP^-}$	Input Referred Negative Trip Point	(see Figure 1)	-4	-8	mV min

## 2.7V Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V_{CM} = V^+/2$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ ,  $C_L = 10\text{pF}$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Limit (Note 6)	Units
$t_{PD}$	Propagation Delay	Overdrive = 5mV $V_{CM} = 0\text{V}$ (Note 7)	12		ns max
		Overdrive = 15mV $V_{CM} = 0\text{V}$ (Note 7)	11		
		Overdrive = 50mV $V_{CM} = 0\text{V}$ (Note 7)	10	20	
$t_{SKEW}$	Propagation Delay Skew	(Note 8)	1		ns
$t_r$	Output Rise Time	10% to 90%	2.5		ns
$t_f$	Output Fall Time	90% to 10%	2		ns

## 5V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V_{CM} = V^+/2$ ,  $V^+ = 5\text{V}$ ,  $V^- = 0\text{V}$ ,  $C_L = 10\text{pF}$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Limit (Note 6)	Units
$V_{OS}$	Input Offset Voltage		1	6 <b>8</b>	mV max
$I_B$	Input Bias Current		500	950 <b>2000</b>	nA max
$I_{OS}$	Input Offset Current		50	200 <b>400</b>	nA max
CMRR	Common Mode Rejection Ratio	$0\text{V} < V_{CM} < 3.8\text{V}$	85	65 <b>55</b>	dB min
PSRR	Power Supply Rejection Ratio	$V^+ = 2.7\text{V}$ to $5\text{V}$	85	65 <b>55</b>	dB min
$V_{CM}$	Input Common-Mode Voltage Range	CMRR > 50dB	$V_{CC} - 1$	$V_{CC} - 1.2$ $V_{CC} - 1.3$	V min
			-0.2	-0.1 0	V max
$V_O$	Output Swing High	$I_L = 4\text{mA}$ , $V_{ID} = 500\text{mV}$	$V_{CC} - 0.13$	$V_{CC} - 0.2$ $V_{CC} - 0.3$	V min
		$I_L = 0.4\text{mA}$ , $V_{ID} = 500\text{mV}$	$V_{CC} - 0.02$	$V_{CC} - 0.05$ $V_{CC} - 0.15$	
	Output Swing Low	$I_L = -4\text{mA}$ , $V_{ID} = -500\text{mV}$	80	180 <b>280</b>	mV max
		$I_L = -0.4\text{mA}$ , $V_{ID} = -500\text{mV}$	10	50 <b>150</b>	
$I_{SC}$	Output Short Circuit Current	Sourcing, $V_O = 0\text{V}$ (Note 3)	68	30 <b>20</b>	mA min
		Sinking, $V_O = 5\text{V}$ (Note 3)	65	30 <b>20</b>	
$I_S$	Supply Current	No load	1.1	1.8 <b>2.4</b>	mA max
$V_{HYST}$	Input Hysteresis Voltage	(Note 10)	7.5		mV
$V_{Trip^+}$	Input Referred Positive Trip Point	(See figure 1)	3.5	8	mV max
$V_{Trip^-}$	Input Referred Negative Trip Point	(See figure 1)	-4	-8	mV min

## 5V Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V_{CM} = V^+/2$ ,  $V^+ = 5\text{V}$ ,  $V^- = 0\text{V}$ ,  $C_L = 10\text{pF}$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Limit (Note 6)	Units
$t_{PD}$	Propagation Delay	Overdrive = 5mV $V_{CM} = 0\text{V}$ (Note 7)	9		ns max
		Overdrive = 15mV $V_{CM} = 0\text{V}$ (Note 7)	8	20	
		Overdrive = 50mV $V_{CM} = 0\text{V}$ (Note 7)	7	19	
$t_{SKEW}$	Propagation Delay Skew	(Note 8)	0.4		ns
$t_r$	Output Rise Time	10% to 90%	1.3		ns
$t_f$	Output Fall Time	90% to 10%	1.25		ns

**Note 1:** Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical characteristics.

**Note 2:** Human body model,  $1.5\text{k}\Omega$  in series with  $100\text{pF}$ . Machine model,  $200\Omega$  in series with  $100\text{pF}$ .

**Note 3:** Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of  $150^\circ\text{C}$ . Output currents in excess of  $\pm 30\text{mA}$  over long term may adversely affect reliability.

**Note 4:** The maximum power dissipation is a function of  $T_{J(max)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(max)} - T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

**Note 5:** Typical Values represent the most likely parametric norm.

**Note 6:** All limits are guaranteed by testing or statistical analysis.

**Note 7:** Propagation delay measurements made with  $100\text{mV}$  steps. Overdrive is measured relative to  $V_{Trip}$ .

**Note 8:** Propagation Delay Skew is defined as absolute value of the difference between  $t_{PD,LH}$  and  $t_{PD,HL}$ .

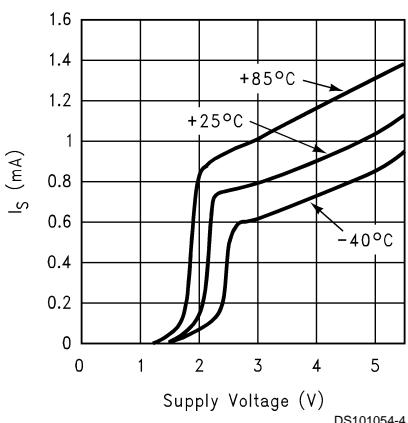
**Note 9:** Limiting input pin current is only necessary for input voltages that exceed absolute maximum input voltage ratings.

**Note 10:** The LMV7219 comparator has internal hysteresis. The trip points are the input voltage needed to change the output state in each direction. The offset voltage is defined as the average of  $V_{trip+}$  and  $V_{trip-}$ , while the hysteresis voltage is the difference of these two.

## Typical Performance Characteristics

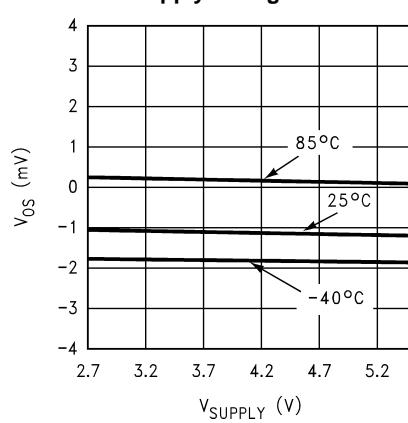
Unless otherwise specified,  $V_S = 5V$ ,  $C_L = 10pF$ ,  $T_A = 25^\circ C$

Supply Current vs.  
Supply Voltage



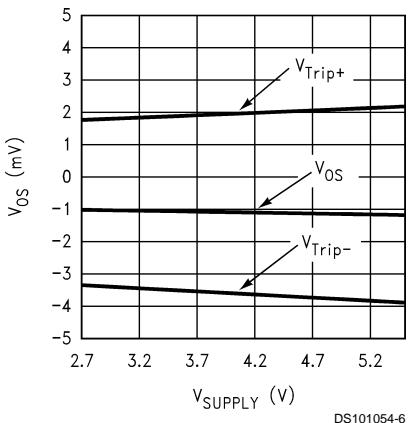
DS101054-4

$V_{OS}$  vs.  
Supply Voltage



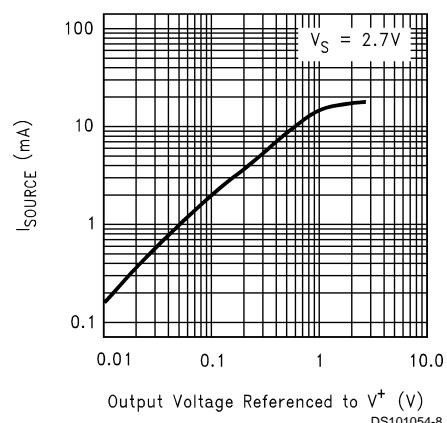
DS101054-5

Input Offset and Trip Voltage vs.  
Supply Voltage



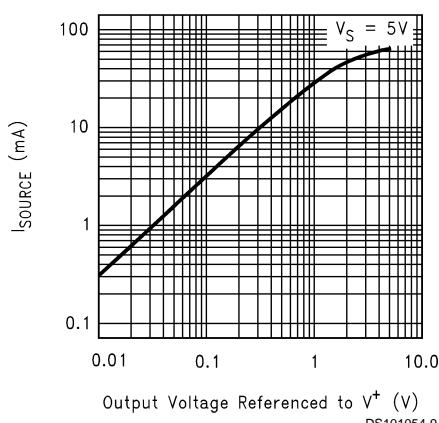
DS101054-6

Sourcing Current vs.  
Output Voltage



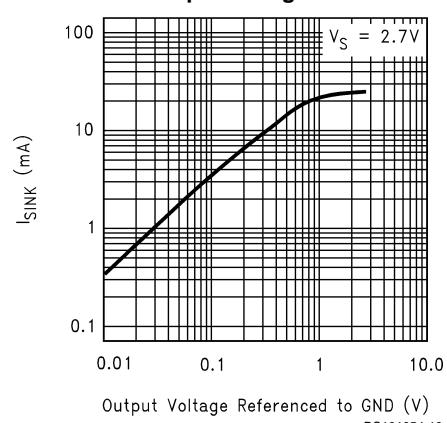
DS101054-8

Sourcing Current vs.  
Output Voltage



DS101054-9

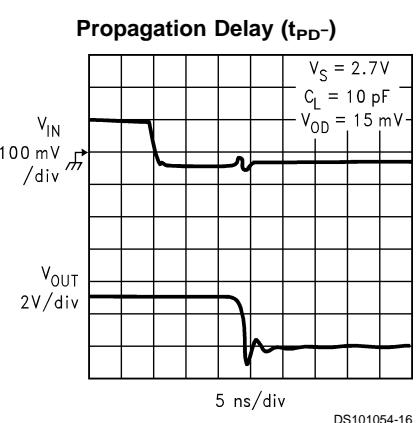
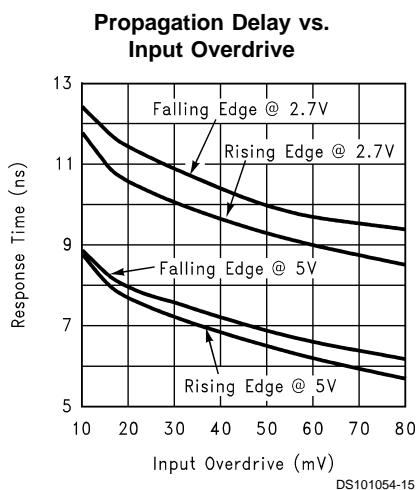
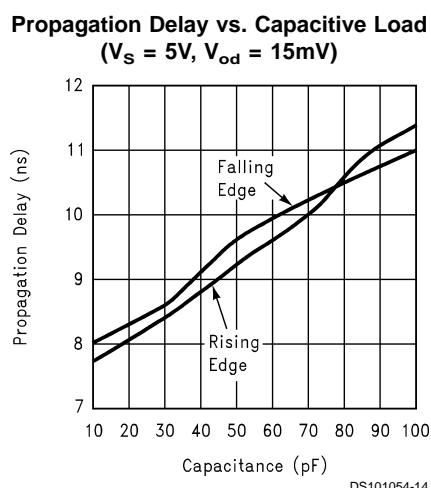
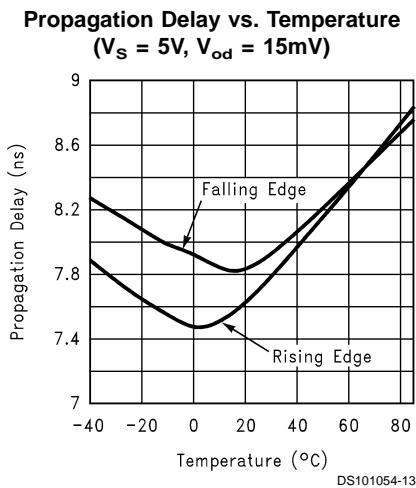
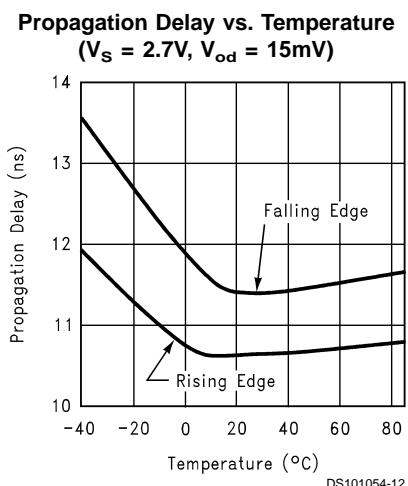
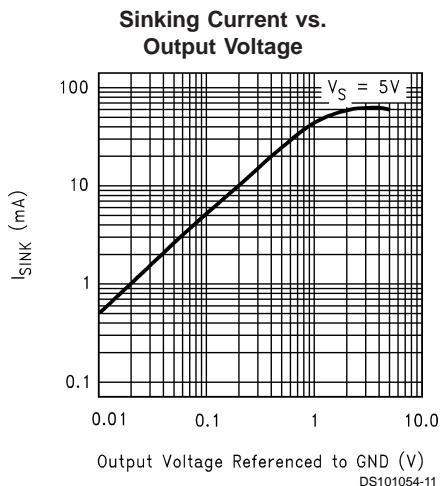
Sinking Current vs.  
Output Voltage



DS101054-10

## Typical Performance Characteristics

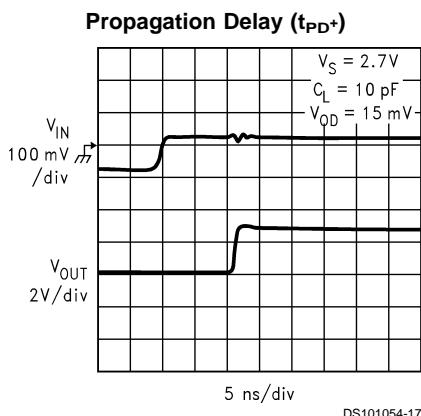
Unless otherwise specified,  $V_S = 5V$ ,  $C_L = 10pF$ ,  $T_A = 25^\circ C$  (Continued)



## Typical Performance Characteristics

## 25°C (Continued)

Unless otherwise specified,  $V_S = 5V$ ,  $C_L = 10pF$ ,  $T_A =$



## Application Section

LMV7219 is a single supply comparator with internal hysteresis, 7ns of propagation delay and only 1.1mA of supply current.

The LMV7219 has a typical input common mode voltage range of  $-0.2V$  below the ground to  $1V$  below  $V_{cc}$ . The differential input stage is a pair of PNP transistors, therefore, the input bias current flows out of the device. If either of the input signals falls below the negative common mode limit, the parasitic PN junction formed by the substrate and the base of the PNP will turn on, resulting in an increase of input bias current.

If one of the inputs goes above the positive common mode limit, the output will still maintain the correct logic level as long as the other input stays within the common mode range. However, the propagation delay will increase. When both inputs are outside the common mode voltage range, current saturation occurs in the input stage, and the output becomes unpredictable.

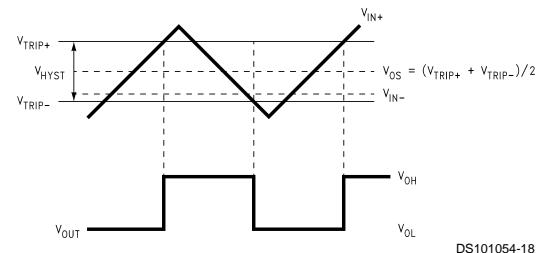
The propagation delay does not increase significantly with large differential input voltages. However, large differential voltages greater than the supply voltage should be avoided to prevent damages to the input stage.

The LMV7219 has a push pull output. When the output switches, there is a direct path between  $V_{cc}$  and ground, causing high output sinking or sourcing current during the transition. After the transition, the output current decreases and the supply current settles back to about 1.1mA at 5V, thus conserving power consumption.

Most high-speed comparators oscillate when the voltage of one of the inputs is close to or equal to the voltage on the other input due to noise or undesirable feedback. The LMV7219 have 7mV of internal hysteresis to counter parasitic effects and noise. The hysteresis does not change significantly with the supply voltages and the common mode input voltages as reflected in the specification table.

The internal hysteresis creates two trip points, one for the rising input voltage and one for the falling input voltage. The difference between the trip points is the hysteresis. With internal hysteresis, when the comparator's input voltages are equal, the hysteresis effectively causes one comparator-input voltage to move quickly past the other, thus taking the input out of the region where oscillation occurs. Standard

comparators require hysteresis to be added with external resistors. The fixed internal hysteresis eliminates these resistors.



**FIGURE 1. Input and Output Waveforms, Non-Inverting Input Varied**

## Additional Hysteresis

If additional hysteresis is desired, this can be done with the addition of three resistors using positive feedback, as shown in *Figure 2*. The positive feedback method slows the comparator response time. Calculate the resistor values as follows:

1) Select R3. The current through R3 should be greater than the input bias current to minimize errors. The current through R3 ( $I_F$ ) at the trip point is  $(V_{REF} - V_{OUT}) / R3$ . Consider the two possible output states when solving for R3, and use the smaller of the two resulting resistor values. The two formulas are:

$$R3 = V_{REF} / I_F \quad (\text{when } V_{OUT} = 0)$$

$$R3 = V_{CC} - V_{REF} / I_F \quad (V_{OUT} = V_{CC})$$

2) Choose a hysteresis band required ( $V_{HB}$ ).

3) Calculate R1, where  $R1 = R3 X (V_{HB} / V_{CC})$

4) Choose the trip point for  $V_{IN}$  rising. This is the threshold voltage ( $V_{THR}$ ) at which the comparator switches from low to high as  $V_{IN}$  rises about the trip point.

5) Calculate R2 as follows:

$$R_2 = \frac{1}{\left(\frac{V_{THR}}{V_{REF} \times R_1}\right) - \frac{1}{R_1} - \frac{1}{R_3}}$$

6) Verify the trip voltage and hysteresis as follows:

## Application Section (Continued)

$$V_{IN} \text{ rising: } V_{THR} = V_{REF} \times R_1 \times \left( \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right)$$

$$V_{IN} \text{ falling: } V_{THF} = V_{THR} - \left( \frac{R_1 \times V_{CC}}{R_3} \right)$$

$$\text{Hysteresis} = V_{THR} - V_{THF}$$

This method is recommended for additional hysteresis of up to a few hundred millivolts. Beyond that, the impedance of R3 is low enough to affect the bias string and adjustment of R1 may be also required.

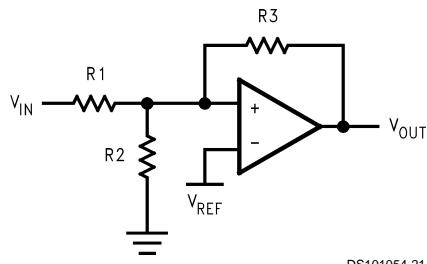


FIGURE 2. Additional Hysteresis

### Circuit Layout and Bypassing

The LMV7219 requires high-speed layout. Follow these layout guidelines:

1. Power supply bypassing is critical, and will improve stability and transient response. A decoupling capacitor such as  $0.1\mu\text{F}$  ceramic should be placed as close as possible to  $V^+$  pin. An additional  $2.2\mu\text{F}$  tantalum capacitor may be required for extra noise reduction.
2. Keep all leads short to reduce stray capacitance and lead inductance. It will also minimize unwanted parasitic feedback around the comparator.
3. The device should be soldered directly to the PC board instead of using a socket.
4. Use a PC board with a good, unbroken low inductance ground plane. Make sure ground paths are low-impedance, especially where heavier currents are flowing.
5. Input traces should be kept away from output traces. This can be achieved by running a topside ground plane between the output and inputs.
6. Run the ground trace under the device up to the bypass capacitor to shield the inputs from the outputs.
7. To prevent parasitic feedback when input signals are slow-moving, a small capacitor of  $1000\text{pF}$  or less can be placed between the inputs. It can also help eliminate oscillations in the transition region. However, this capacitor can cause some degradation to tpd when the source impedance is low.

### Zero-Crossing Detector

The inverting input is connected to ground and the non-inverting input is connected to  $100\text{mVp-p}$  signal. As the signal at the non-inverting input crosses  $0\text{V}$ , the comparator's output Changes State.

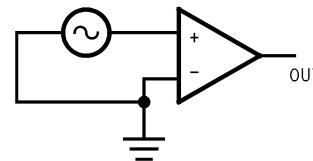


FIGURE 3. Zero-Crossing Detector

### Threshold Detector

Instead of tying the inverting input to  $0\text{V}$ , the inverting input can be tied to a reference voltage. The non-inverting input is connected to the input. As the input passes the  $V_{REF}$  threshold, the comparator's output changes state.

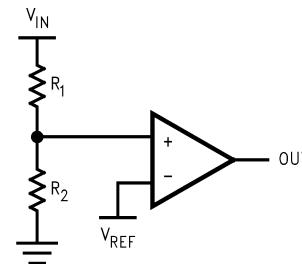


FIGURE 4. Threshold Detector

### Crystal Oscillator

A simple crystal oscillator using the LMV7219 is shown below. Resistors R1 and R2 set the bias point at the comparator's non-inverting input. Resistors R3, R4 and C1 sets the inverting input node at an appropriate DC average level based on the output. The crystal's path provides resonant positive feedback and stable oscillation occurs. The output duty cycle for this circuit is roughly 50%, but it is affected by resistor tolerances and to a lesser extent by the comparator offset.

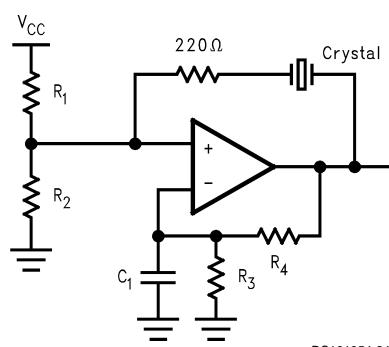
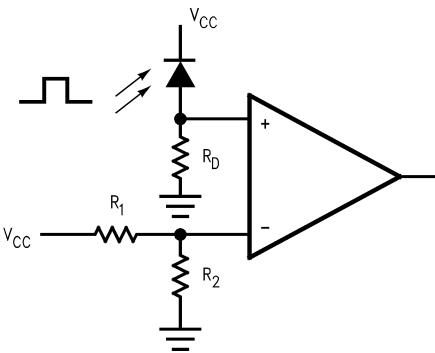


FIGURE 5. Crystal Oscillator

## Application Section (Continued)

### IR Receiver

The LMV7219 is an ideal candidate to be used as an infrared receiver. The infrared photo diode creates a current relative to the amount of infrared light present. The current creates a voltage across RD. When this voltage level cross the voltage applied by the voltage divider to the inverting input, the output transitions.

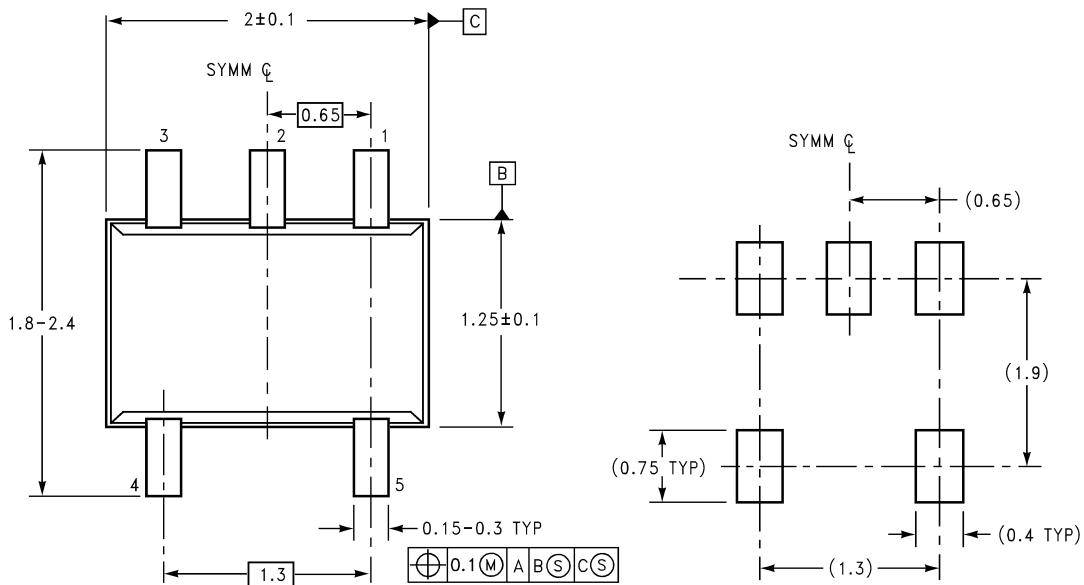


DS101054-25

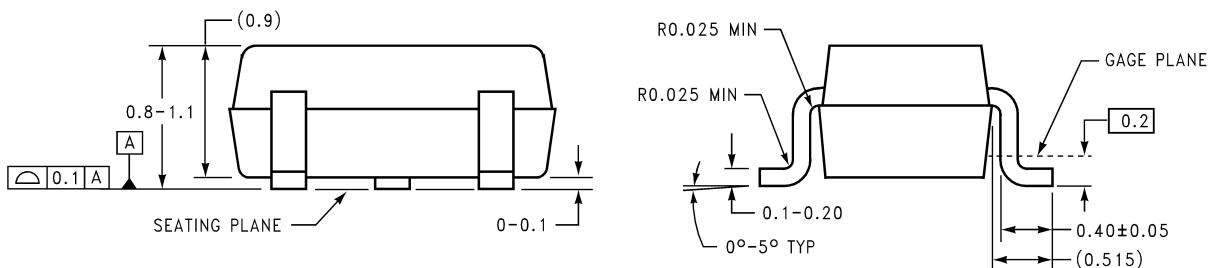
FIGURE 6. IR Receiver

## Physical Dimensions

inches (millimeters) unless otherwise noted



#### LAND PATTERN RECOMMENDATION



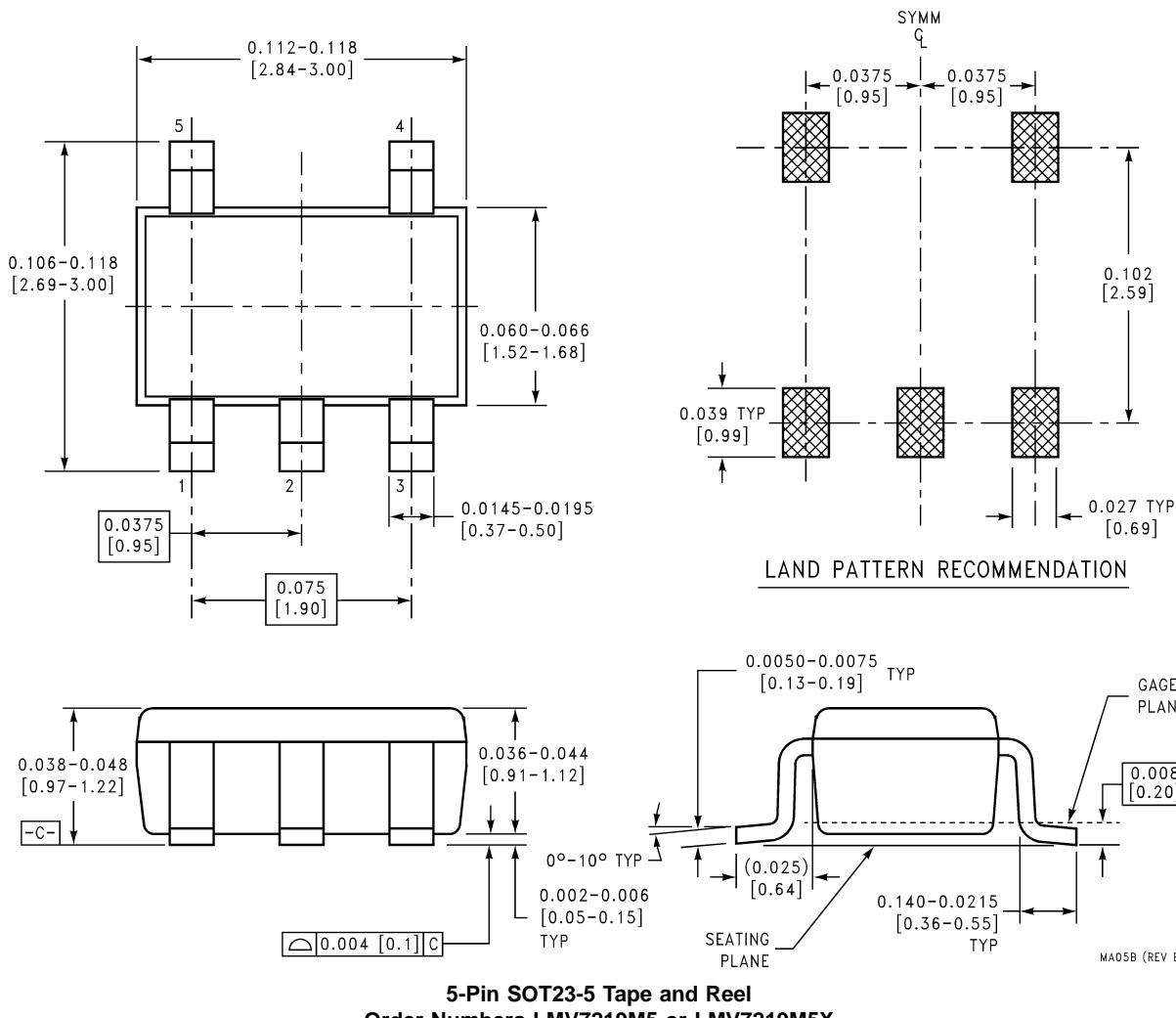
DIMENSIONS ARE IN MILLIMETERS

MAA05A (REV B)

**5-Pin SC70-5 Tape and Reel**  
**Order Numbers LMV7219M7 or LMV7219M7X**  
**NS Package Number MAA05A**

# LMV7219 7 nsec, 2.7V to 5V Comparator with Rail-to Rail Output

## Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



### 5-Pin SOT23-5 Tape and Reel

Order Numbers LMV7219M5 or LMV7219M5X

NS Package Number MA05B

## LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT AND GENERAL COUNSEL OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



**National Semiconductor Corporation**  
Americas  
Tel: 1-800-272-9959  
Fax: 1-800-737-7018  
Email: support@nsc.com  
www.national.com

**National Semiconductor Europe**  
Fax: +49 (0) 180-530 85 86  
Email: europe.support@nsc.com  
Deutsch Tel: +49 (0) 69 9508 6208  
English Tel: +44 (0) 870 24 0 2171  
Français Tel: +33 (0) 1 41 91 8790

**National Semiconductor Asia Pacific Customer Response Group**  
Tel: 65-2544466  
Fax: 65-2504466  
Email: ap.support@nsc.com

**National Semiconductor Japan Ltd.**  
Tel: 81-3-5639-7560  
Fax: 81-3-5639-7507